

Reflectarray-based Intelligent Reflecting Surface to Improve mm-Wave 5G coverage in outdoor scenarios

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Abstract—This work presents the use of a reflectarray panel as an efficient and low-cost solution to improve millimeter-wave coverage in 5G scenarios. A dual-polarized shaped-beam is designed to radiate a flat-top beam to cover a blind zone without coverage. The design is carried out using a phase-only synthesis considering the beam specifications for the outdoor scenario. Then, a single-layer reflectarray element is used to perform a design at 27.7 GHz. This element provides an angularly-stable phase response at the operational frequency with an independent phase control in each linear polarization. The reflectarray panel shows a stable gain and a low both side-lobe level and cross-polarization within 1 GHz bandwidth.

I. INTRODUCTION

The arrival of the Fifth Generation (5G) of mobile systems has sparked the use of millimeter-wave (mm-wave) spectrum to deliver high-speed wireless access in cellular networks. 5G mm-wave bands (mainly at 28, 39, and 60 GHz) aim to reach higher data rates, lower latency, and higher system capacities than the previous generation deployed at frequencies lower than 6 GHz (sub-6 GHz band). A major challenge of mm-wave communications is related to the signal propagation, which is more adverse than at the frequencies used within sub-6GHz communications. The propagation at mm-waves implies a significant increase of the path losses as well as larger penetration losses through obstacles, thus a higher sensitivity to physical barriers [1]. Throughout the last years, Intelligent Reflecting Surfaces (IRS) have been proposed to overcome both issues. An IRS is comprised of a planar arrangement of reflecting elements, which are designed to independently control the amplitude and/or phase of the reflected wave.

In this work, a reflectarray antenna is proposed as a low-cost alternative for passive IRS to improve wireless coverage for 5G mm-wave networks at 28 GHz. The IRS is presented to generate a shaped beam operating in dual polarization. The design is carried out using a phase-only synthesis aiming to concentrate the power received from the base station. The reflectarray is designed using a dipole-based element that provides independent control of both linear polarizations. The reflectarray shows a stable response within the band of 27.2 to

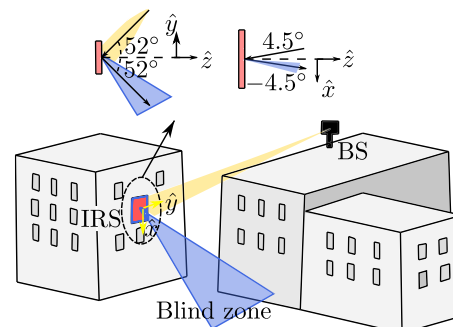


Figure 1. Sketch of an outdoor scenario with an IRS installed on the wall and illuminated from a FR2 5G base station (BS).

28.2 GHz, despite having high angles of incidence in the field provided by the base station.

II. 5G MM-WAVE SCENARIO

The proposed reflectarray-based IRS is designed to be placed in an outdoor scenario to enhance the coverage within a blind zone, as the one shown in Fig. 1. The distance between the base station (BS) and the IRS is around 43.8 m with an angle of 52° and 4.5° in azimuth and elevation, respectively. The outgoing beam is reflected in the specular direction in azimuth and -4.5° in elevation. Besides, the half-power beamwidth (HPBW) is 5.5° in azimuth and 1.5° in elevation. These specifications provide proper coverage within the blind zone. The IRS is comprised of 178×90 elements distributed in a regular grid of $5 \text{ mm} \times 4.1 \text{ mm}$, with an aperture size of $800 \text{ mm} \times 400 \text{ mm}$. The extremely large dimension between the BS and the IRS leads to an average value of the incidence angles of $\theta_i = 51.9^\circ$ and $\varphi_i = 84.5^\circ$, which implies a more complex design process.

III. IRS DESIGN AND VALIDATION

As a first approach, the phasing elements are designed to deflect the beam in a specific direction producing a pencil beam.

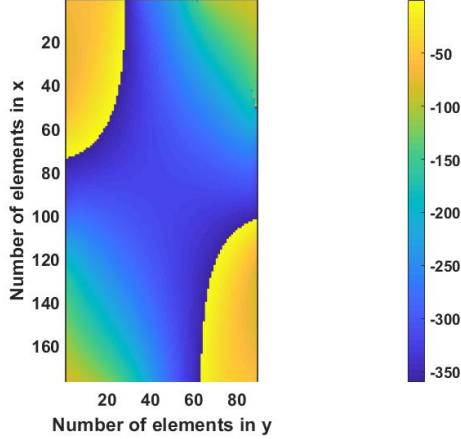


Figure 2. Phase distribution (in degrees) after the phase-only synthesis.

To collimate the beam in a direction (θ_a, φ_a) , the phase-shift of the element can be computed as

$$\phi_{IRS}(x_m, y_n) = k_0(d_{BS} - (x_m \cos \varphi_a + y_n \sin \varphi_a) \sin \theta_a) \quad (1)$$

where k_0 is the wave-number; d_{BS} is the distance from the BS to the (m, n) th element. In this case, (θ_a, φ_a) points to the center of the blind zone and the operational frequency is 27.7 GHz. Note that, the phase-shift of the IRS elements should be identical for both polarizations to ensure that the polarization of the reflected beam does not change.

The phase-shift obtained with (1) generates a beam with the highest gain in the desired direction. However, it presents a HPBW around 2° in azimuth and 0.7° , which is not wide enough to cover the blind zone. To enhance the beam performance a phase-only synthesis is carried out using the generalized Intersection Approach (IA) [2]. The synthesis aims to reach a flat-top beam with a maximum ripple limited to 3 dB. The HPBW is set to satisfy the coverage performance (5.5° in azimuth and 1.5° in elevation). After this process, the phasing element distribution of Fig. 2 is achieved, whose radiation pattern is shown in Fig. 3 (dotted line). Then, this phase distribution is used to carry out a design using an element based on dipoles [3].

The printed elements of the IRS are designed to provide the phase-shift distribution obtained with IA for each polarization. The design process is performed element-by-element adjusting the dipoles length to provide the required phase-shift in both polarizations. The radiation pattern of the resulting layout is analyzed by electromagnetic simulation, using an in-house code based on the Method of Moment (MoM) and local periodicity approach [3]. The azimuth and elevation cuts of the simulated radiation patterns in V and H polarizations are compared at 27.7 GHz, Fig. 3. As can be seen, the beam pointing (51° in azimuth and 18° in elevation) and the HPBW (5.5° in azimuth and 1.5° in elevation) are both compliant with the specifications. The cross-polar component is also analyzed, being 20 dB lower than the co-polar one within the region of the main lobe. Similar performances are found for both linear polarizations within a 1 GHz bandwidth.

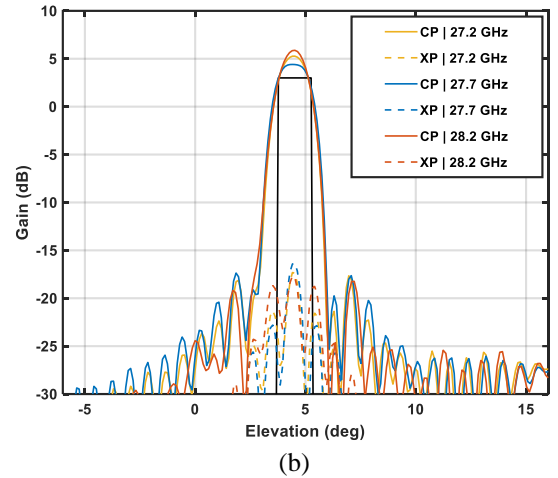
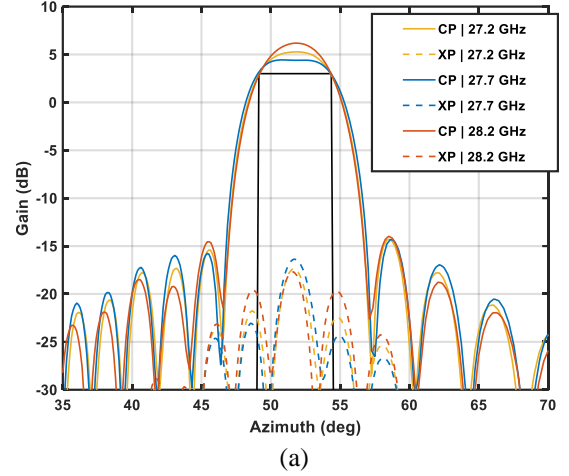


Figure 3. Simulated radiation pattern in V-polarization at the central and extreme frequencies in (a) azimuth and (b) elevation, for the co-copolar (CP) and cross-copolar (XP) components.

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